

CTFD – Frequency Response (7A)

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Computing a Transfer Function

$$Q(D) \cdot y(t) = P(D) \cdot x(t)$$

ZSR

$$Q(s) \cdot Y(s) = P(s) \cdot X(s)$$

$$H(s) = \frac{Y(s)}{X(s)} = \frac{P(s)}{Q(s)}$$

Transfer function (s-domain)

$$H(s) = \frac{Y(s)}{X(s)}$$

Polynomials of Differential Equation

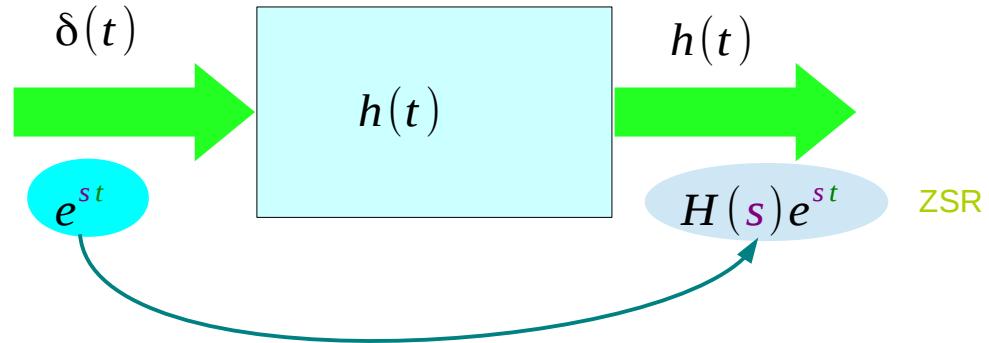
$$H(s) = \frac{P(s)}{Q(s)}$$

$$(b_0, b_1, \dots, b_{N-1}, b_N)$$

$$(1, a_1, \dots, a_{N-1}, a_N)$$

input applied
at $t = -\infty$

zero initial conditions



Transfer function (t -domain)

$$H(s) = \left[\frac{y(t)}{x(t)} \right]_{x(t)=e^s}$$

Laplace Transform of $h(t)$

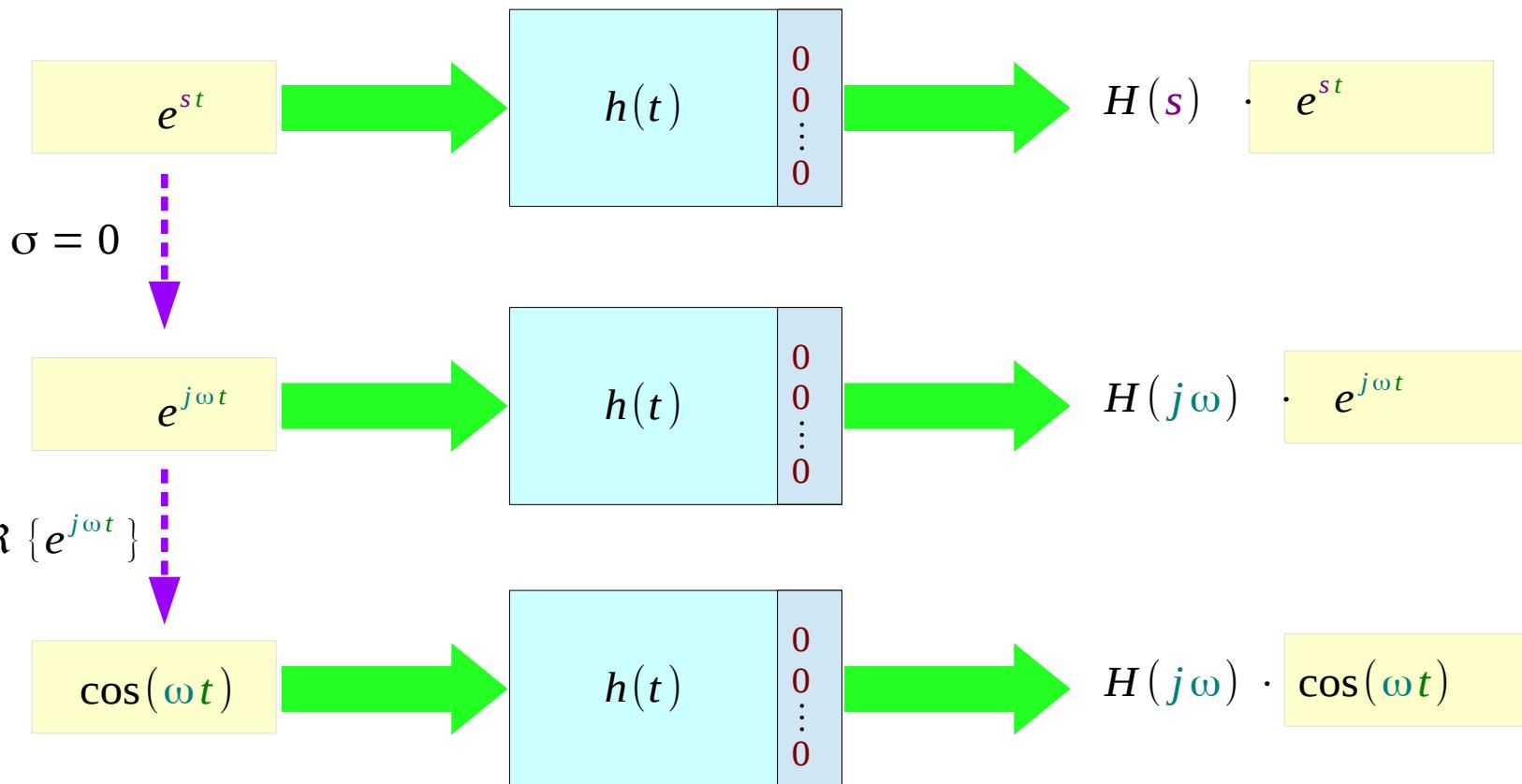
$$H(s) = \int_0^{+\infty} h(\tau) e^{-s\tau} d\tau$$

Special cases of H(s)

$$s = \sigma + j\omega$$

input applied
at $t = -\infty$

general
cases



Frequency Response

Laplace Transform of $h(t)$

$$H(s) = \int_{-\infty}^{+\infty} h(\tau) e^{-s\tau} d\tau \quad s = \sigma + j\omega$$

Fourier Transform of $h(t)$

$$H(j\omega) = \int_{-\infty}^{+\infty} h(\tau) e^{-j\omega\tau} d\tau \quad s = j\omega$$

Polynomials of Differential Equation

$$H(s) = \frac{P(s)}{Q(s)} \quad s = \sigma + j\omega$$

Polynomials of Differential Equation

$$H(j\omega) = \frac{P(j\omega)}{Q(j\omega)} \quad s = j\omega$$

Transfer function (t-domain)

$$H(s) = \left[\frac{y(t)}{x(t)} \right]_{x(t)=e^{st}} \quad s = \sigma + j\omega$$

Frequency response (t-domain)

$$H(j\omega) = \left[\frac{y(t)}{x(t)} \right]_{x(t)=e^{j\omega t}} \quad s = j\omega$$

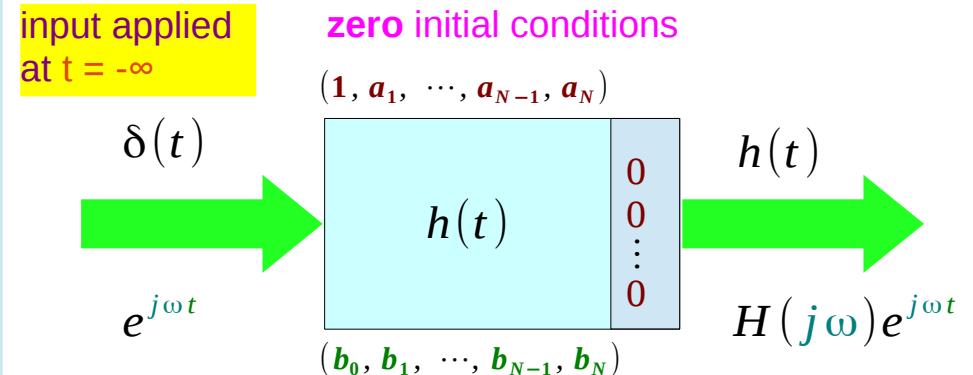
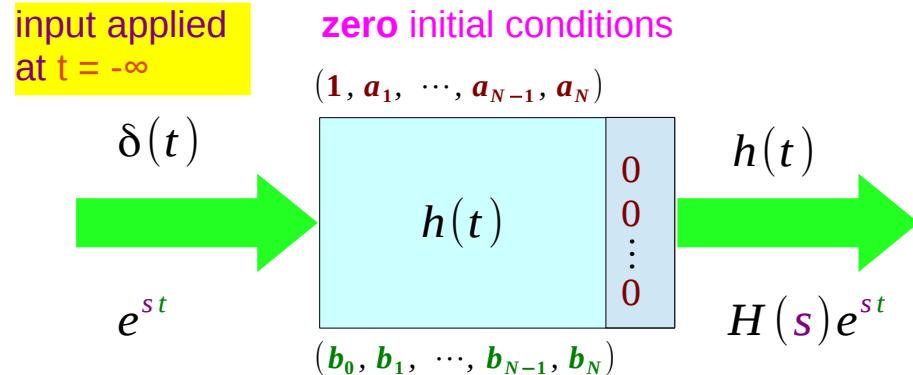
Transfer function (s-domain)

$$H(s) = \frac{Y(s)}{X(s)} \quad s = \sigma + j\omega$$

Frequency response (ω -domain)

$$H(j\omega) = \frac{Y(j\omega)}{X(j\omega)} \quad s = j\omega$$

Transfer Function & Frequency Response



$$\begin{aligned}
 y(t) &= h(t) * e^{st} \quad s = \sigma + j\omega \\
 &= \int_{-\infty}^{+\infty} h(\tau) e^{s(t-\tau)} d\tau \\
 &= e^{st} \cdot \boxed{\int_{-\infty}^{+\infty} h(\tau) e^{-s\tau} d\tau} \\
 &= e^{st} \cdot \boxed{H(s)}
 \end{aligned}$$

$$h(t) = 0 \quad (t < 0)$$

$$\begin{aligned}
 y(t) &= h(t) * e^{j\omega t} \quad s = j\omega \\
 &= \int_{-\infty}^{+\infty} h(\tau) e^{j\omega(t-\tau)} d\tau \\
 &= e^{j\omega t} \cdot \boxed{\int_{-\infty}^{+\infty} h(\tau) e^{-j\omega\tau} d\tau} \\
 &= e^{j\omega t} \cdot \boxed{H(j\omega)}
 \end{aligned}$$

$$h(t) = 0 \quad (t < 0)$$

$$H(s) = \int_0^{+\infty} h(\tau) e^{-s\tau} d\tau$$

Transfer function

$$H(j\omega) = \int_{-\infty}^{+\infty} h(\tau) e^{-j\omega\tau} d\tau$$

Frequency response

Total Response to everlasting sinusoidal inputs

$$e^{+j\omega t} \rightarrow h(t) \begin{pmatrix} 0 \\ 0 \\ \vdots \\ 0 \end{pmatrix} \rightarrow H(+j\omega)e^{+j\omega t} = |H(+j\omega)|e^{+j\arg\{H(+j\omega)\}} \cdot e^{+j\omega t}$$

$$e^{-j\omega t} \rightarrow h(t) \begin{pmatrix} 0 \\ 0 \\ \vdots \\ 0 \end{pmatrix} \rightarrow H(-j\omega)e^{-j\omega t} = |H(-j\omega)|e^{+j\arg\{H(-j\omega)\}} \cdot e^{-j\omega t}$$

$$e^{+j\omega t} + e^{-j\omega t} \rightarrow h(t) \begin{pmatrix} 0 \\ 0 \\ \vdots \\ 0 \end{pmatrix} \rightarrow |H(+j\omega)|e^{[+j\omega t + \arg\{H(+j\omega)\}]} + |H(-j\omega)|e^{[-j\omega t + \arg\{H(-j\omega)\}]}$$

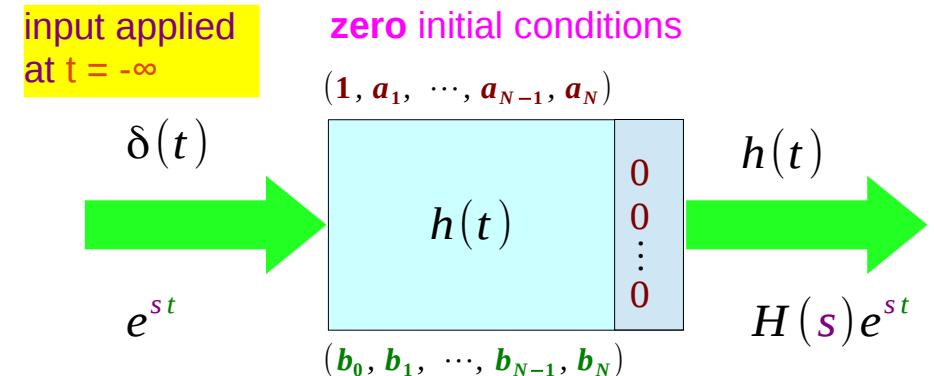
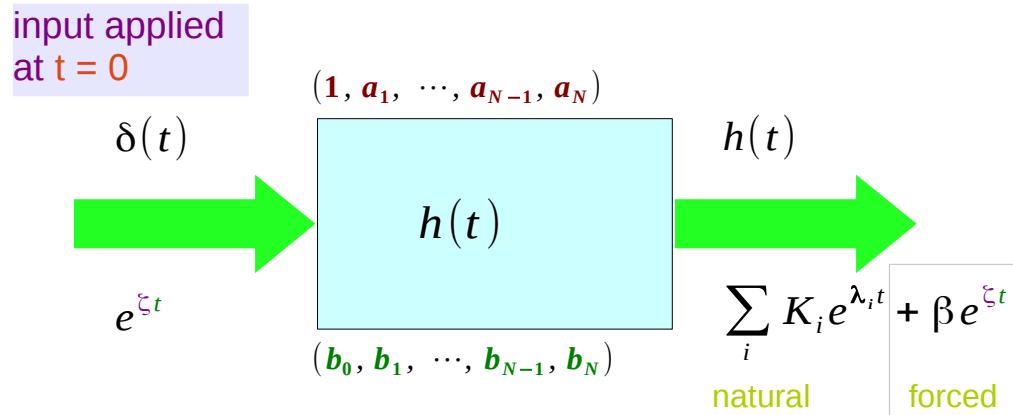
$$= |H(+j\omega)| \{ e^{[+j\omega t + \arg\{H(+j\omega)\}]} + e^{[-j\omega t - \arg\{H(+j\omega)\}]} \}$$

$$2\cos(\omega t) \rightarrow h(t) \begin{pmatrix} 0 \\ 0 \\ \vdots \\ 0 \end{pmatrix} \rightarrow |H(j\omega)| 2\cos(\omega t + \arg\{H(j\omega)\})$$

$$A\cos(\omega t + \alpha) \rightarrow h(t) \begin{pmatrix} 0 \\ 0 \\ \vdots \\ 0 \end{pmatrix} \rightarrow A |H(j\omega)| \cos(\omega t + \alpha + \arg\{H(j\omega)\})$$

$$A\sin(\omega t + \alpha) \rightarrow h(t) \begin{pmatrix} 0 \\ 0 \\ \vdots \\ 0 \end{pmatrix} \rightarrow A |H(j\omega)| \sin(\omega t + \alpha + \arg\{H(j\omega)\})$$

Sinusoidal Steady State Response (1)



total response

natural forced

$$y(t) = \sum_i K_i e^{\lambda_i t} + H(\zeta) e^{\zeta t} \quad t \geq 0$$

$$y(t) = \sum_i K_i e^{\lambda_i t} + H(\zeta) x(t) \quad x(t) = e^{\zeta t}$$

$$Y(s) = \sum_i \frac{K_i}{(s - \lambda_i)} + H(s) X(s)$$

steady state response

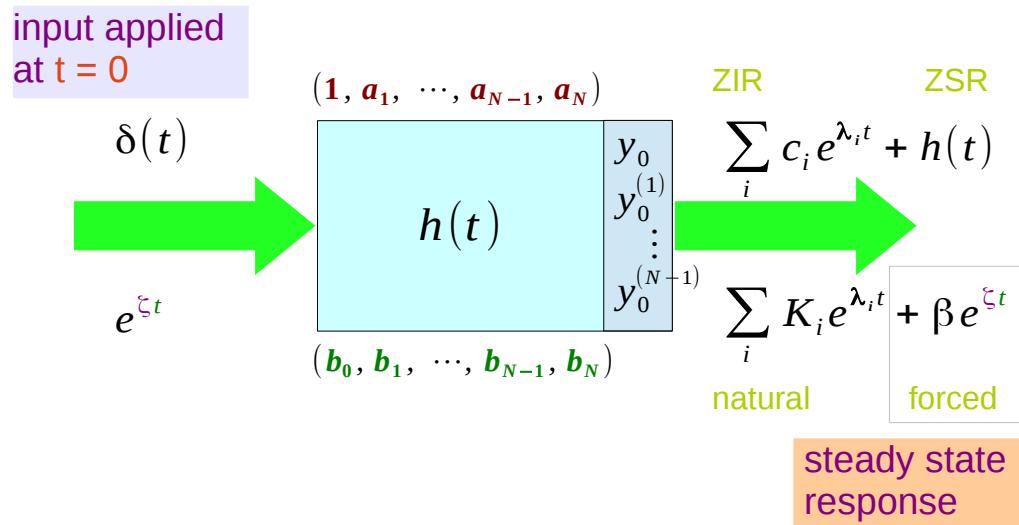
forced

$$y_{ss}(t) = H(\zeta) e^{\zeta t} \quad t \rightarrow \infty$$

$$y_{ss}(t) = H(\zeta) x(t) \quad x(t) = e^{\zeta t}$$

$$Y_{ss}(s) = H(s) X(s)$$

Sinusoidal Steady State Response (2)



$$x(t) = A$$

$$\xi = 0$$

$$h(t) \begin{pmatrix} y_0^{(0)} \\ y_0^{(1)} \\ \vdots \\ y_0^{(N-1)} \end{pmatrix}$$

$$y_{ss}(t) = H(0) \cdot A e^{0t}$$

$$= A \cdot H(0)$$

$$x(t) = A e^{j\omega t}$$

$$\xi = j\omega$$

$$h(t) \begin{pmatrix} y_0^{(0)} \\ y_0^{(1)} \\ \vdots \\ y_0^{(N-1)} \end{pmatrix}$$

$$y_{ss}(t) = H(j\omega) \cdot A e^{j\omega t}$$

$$= A \cdot H(j\omega) e^{j\omega t}$$

$$x(t) = A \cos(\omega t)$$

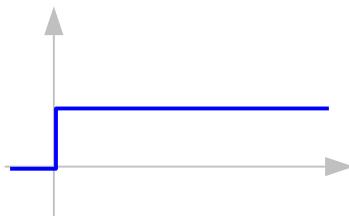
$$\xi = j\omega$$

$$h(t) \begin{pmatrix} y_0^{(0)} \\ y_0^{(1)} \\ \vdots \\ y_0^{(N-1)} \end{pmatrix}$$

$$y_{ss}(t) = H(j\omega) \cdot A \cos(\omega t)$$

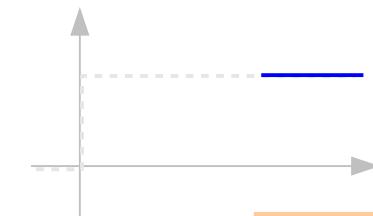
$$= A \cdot H(j\omega) \cos(\omega t)$$

Sinusoidal Steady State Response (3)



$$y_{ss}(t) = H(0) \cdot A e^{0t} \\ = A \cdot H(0)$$

$$x(t) = A \\ \xi = 0$$

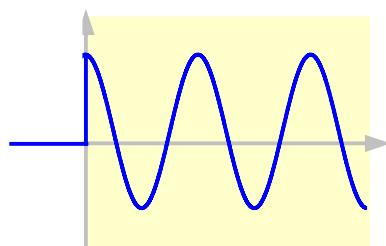
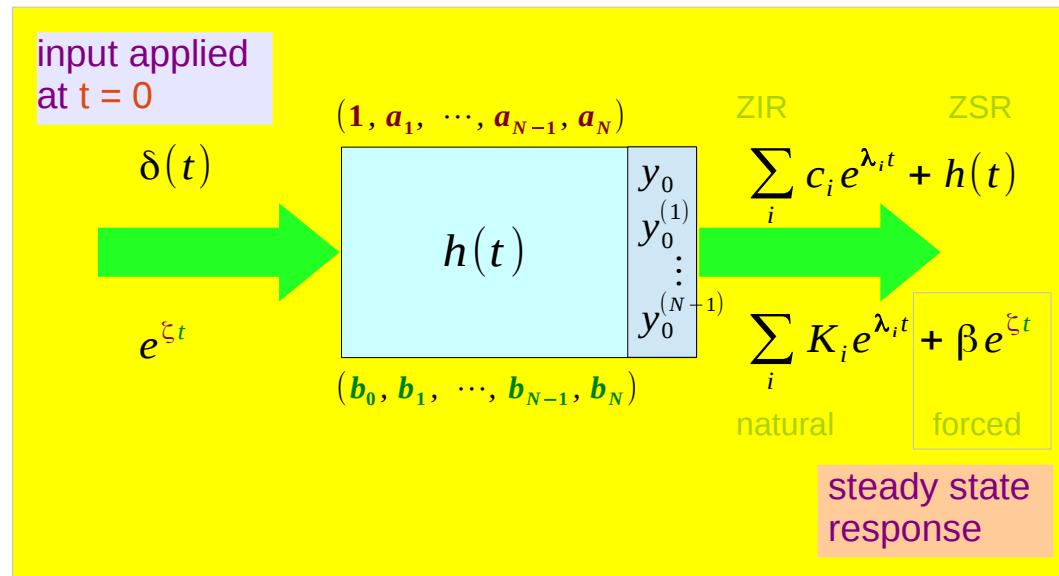


steady state
response

Forced
response

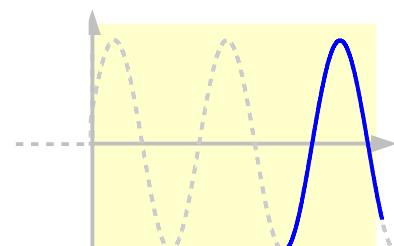
Forced
response

steady state
response



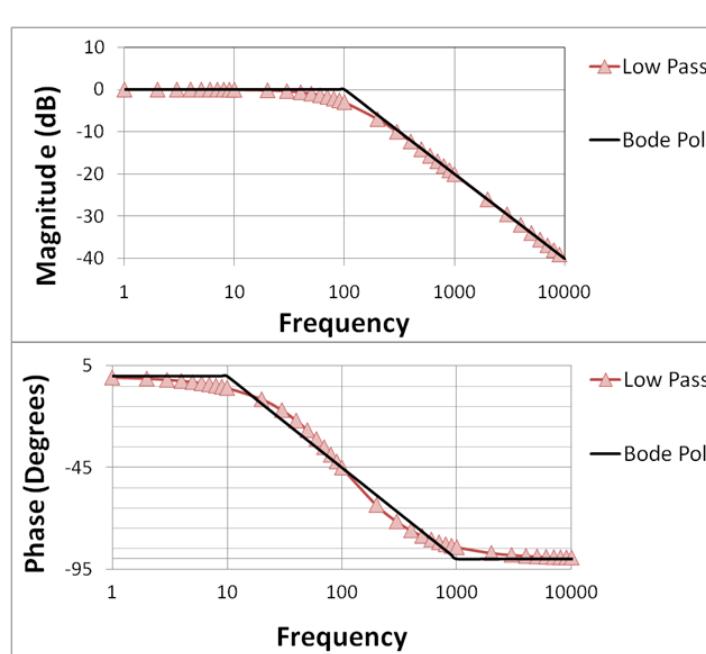
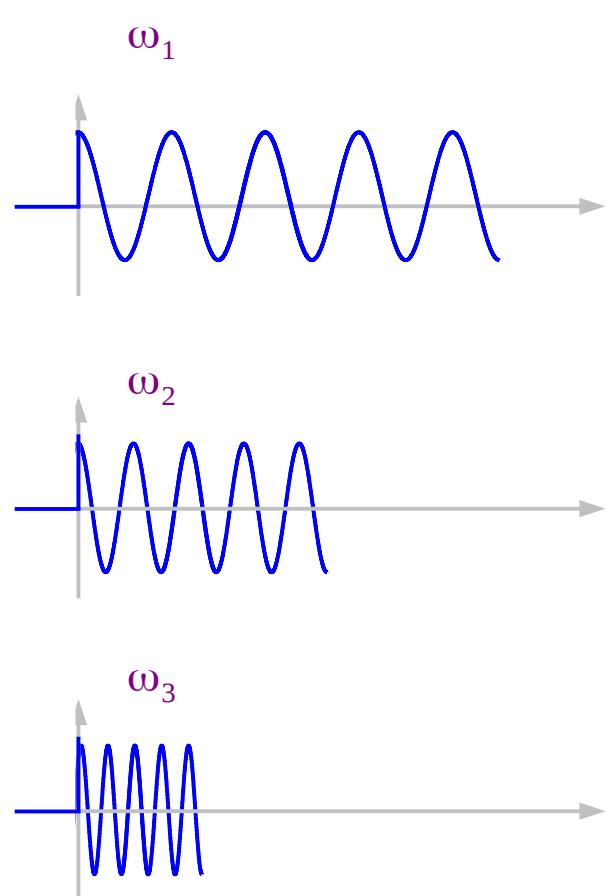
$$y_{ss}(t) = H(j\omega) \cdot A \cos(\omega t) \\ = A \cdot H(j\omega) \cos(\omega t)$$

$$x(t) = A \cos(\omega t) \\ \xi = j\omega$$

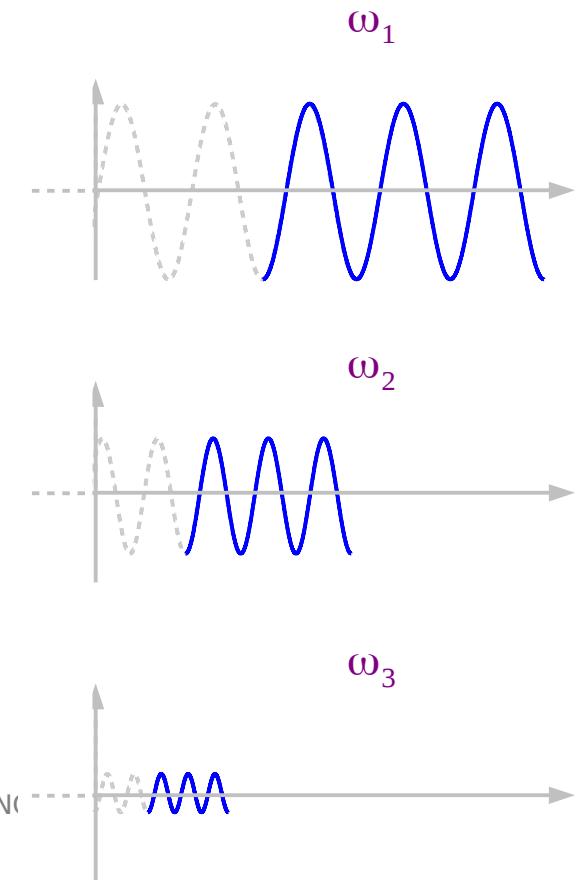


Frequency Response

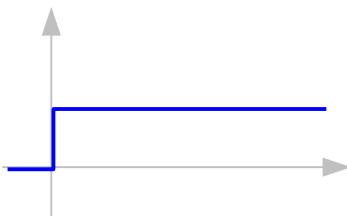
$$\begin{aligned}y_{ss}(t) &= H(j\omega) \cdot A \cos(\omega t) \\&= A \cdot H(j\omega) \cos(\omega t)\end{aligned}$$
$$\xi = j\omega$$



http://en.wikipedia.org/wiki/File:Bode_Low-Pass.PNG



Transient Response



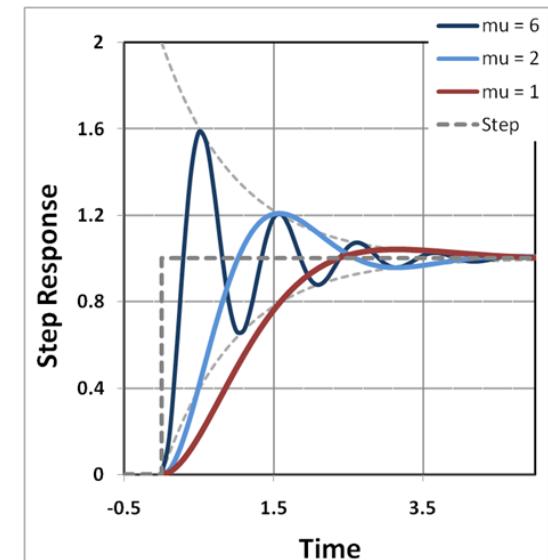
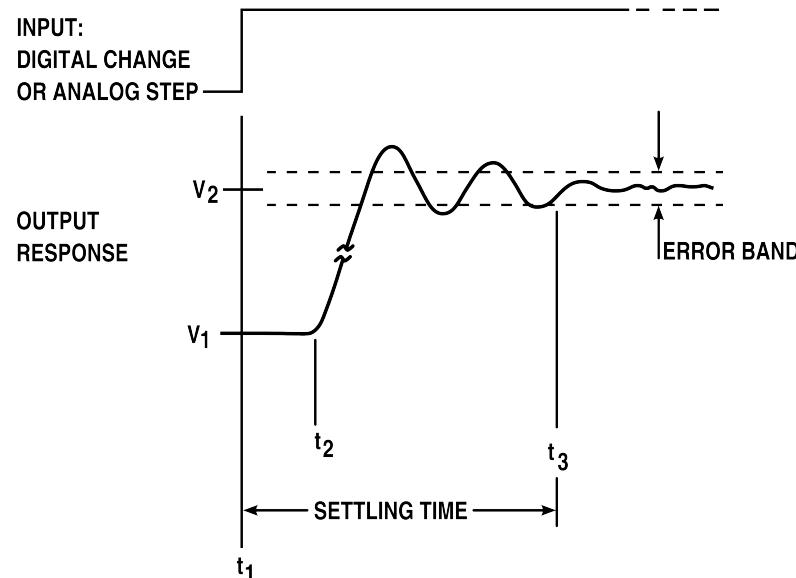
$$y_{ss}(t) = H(0) \cdot A e^{0t} \\ = A \cdot H(0)$$

$$x(t) = A \\ \xi = 0$$



Natural + Forced response

transient response



http://en.wikipedia.org/wiki/File:High_accuracy_settling_time_measurements_figure_1.png
http://en.wikipedia.org/wiki/File:Step_response_for_two-pole_feedback_amplifier.PNG

Frequency Response in Control Theory (1)

$$\sqrt{A^2+B^2} \cos(\omega t - \theta) \rightarrow G(s) \rightarrow |G(j\omega)| \sqrt{A^2+B^2} \cos(\omega t - \theta + \arg\{G(j\omega)\})$$

$$A \cos(\omega t) + B \sin(\omega t)$$

$$= \sqrt{A^2+B^2} \left[\frac{A}{\sqrt{A^2+B^2}} \cos(\omega t) + \frac{B}{\sqrt{A^2+B^2}} \sin(\omega t) \right]$$

$$= \sqrt{A^2+B^2} [\cos(\theta) \cos(\omega t) + \sin(\theta) \sin(\omega t)]$$

$$= \sqrt{A^2+B^2} \cos(\theta - \omega t)$$

$$= \sqrt{A^2+B^2} \cos(\omega t - \theta)$$

$$A \cos(\omega t) + B \sin(\omega t)$$

$$= \sqrt{A^2+B^2} \cos(\omega t - \theta)$$

$$\cos(\theta) = \frac{A}{\sqrt{A^2+B^2}}$$

$$\sin(\theta) = \frac{B}{\sqrt{A^2+B^2}}$$

Frequency Response in Control Theory (2)

$$\sqrt{A^2+B^2} \cos(\omega t - \theta) \rightarrow G(s) \rightarrow |G(j\omega)| \sqrt{A^2+B^2} \cos(\omega t - \theta + \arg\{G(j\omega)\})$$

$$A \cos(\omega t) + B \sin(\omega t) \\ = \sqrt{A^2+B^2} \cos(\omega t - \theta)$$

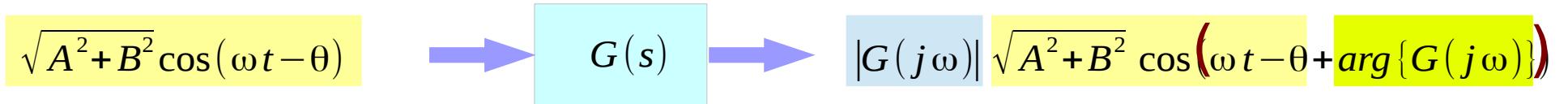
$$\frac{As}{s^2+\omega^2} + \frac{B\omega}{s^2+\omega^2} = \frac{A s + B \omega}{s^2+\omega^2}$$

$$Y(s) = \frac{A s + B \omega}{s^2 + \omega^2} G(s) = \frac{A s + B \omega}{(s + j\omega)(s - j\omega)} G(s) = \frac{K_1}{s + j\omega} + \frac{K_2}{s - j\omega} + F(s) \quad \text{Partial Fraction}$$

$$K_1 = \left[\frac{As+B\omega}{s^2+\omega^2} (s+j\omega) G(s) \right]_{s=-j\omega} = \left[\frac{As+B\omega}{(s-j\omega)} G(s) \right]_{s=-j\omega} = \frac{-A j\omega + B\omega}{-2 j\omega} G(-j\omega) = \frac{1}{2} (A + jB) G(-j\omega)$$

$$K_2 = \left[\frac{As+B\omega}{s^2+\omega^2} (s-j\omega) G(s) \right]_{s=+j\omega} = \left[\frac{As+B\omega}{(s+j\omega)} G(s) \right]_{s=+j\omega} = \frac{A j\omega + B\omega}{+2 j\omega} G(+j\omega) = \frac{1}{2} (A - jB) G(+j\omega)$$

Frequency Response in Control Theory (3)



$$Y(s) = \frac{K_1}{s+j\omega} + \frac{K_2}{s-j\omega} + F(s)$$

$$K_1 = \frac{1}{2}(A+jB)G(-j\omega)$$

$$K_2 = \frac{1}{2}(A-jB)G(+j\omega)$$

$$\begin{aligned} A \pm jB &= \sqrt{A^2+B^2} \left[\frac{A}{\sqrt{A^2+B^2}} \pm j \frac{B}{\sqrt{A^2+B^2}} \right] \\ &= \sqrt{A^2+B^2} [\cos \theta \pm j \sin \theta] \\ &= \sqrt{A^2+B^2} e^{\pm j\theta} \end{aligned}$$

Stable System *system modes*
 $e^{p_i} \rightarrow 0$ $e^{\sigma_i} \rightarrow 0$ $p_i < 0, \sigma_i < 0$

$$\lim_{t \rightarrow \infty} f(t) = 0 \quad \lim_{s \rightarrow \infty} s F(s) = 0$$

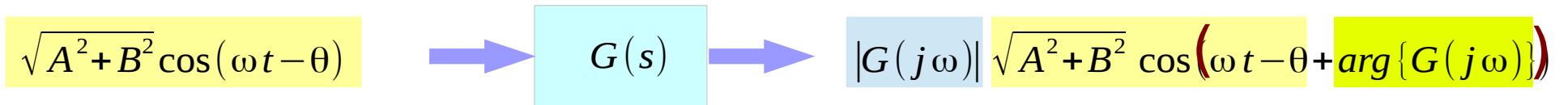
Ignore $F(s)$

$$Y_{ss}(s) = \frac{K_1}{s+j\omega} + \frac{K_2}{s-j\omega}$$

$$K_1 = \frac{1}{2} \sqrt{A^2+B^2} e^{+j\theta} G(-j\omega)$$

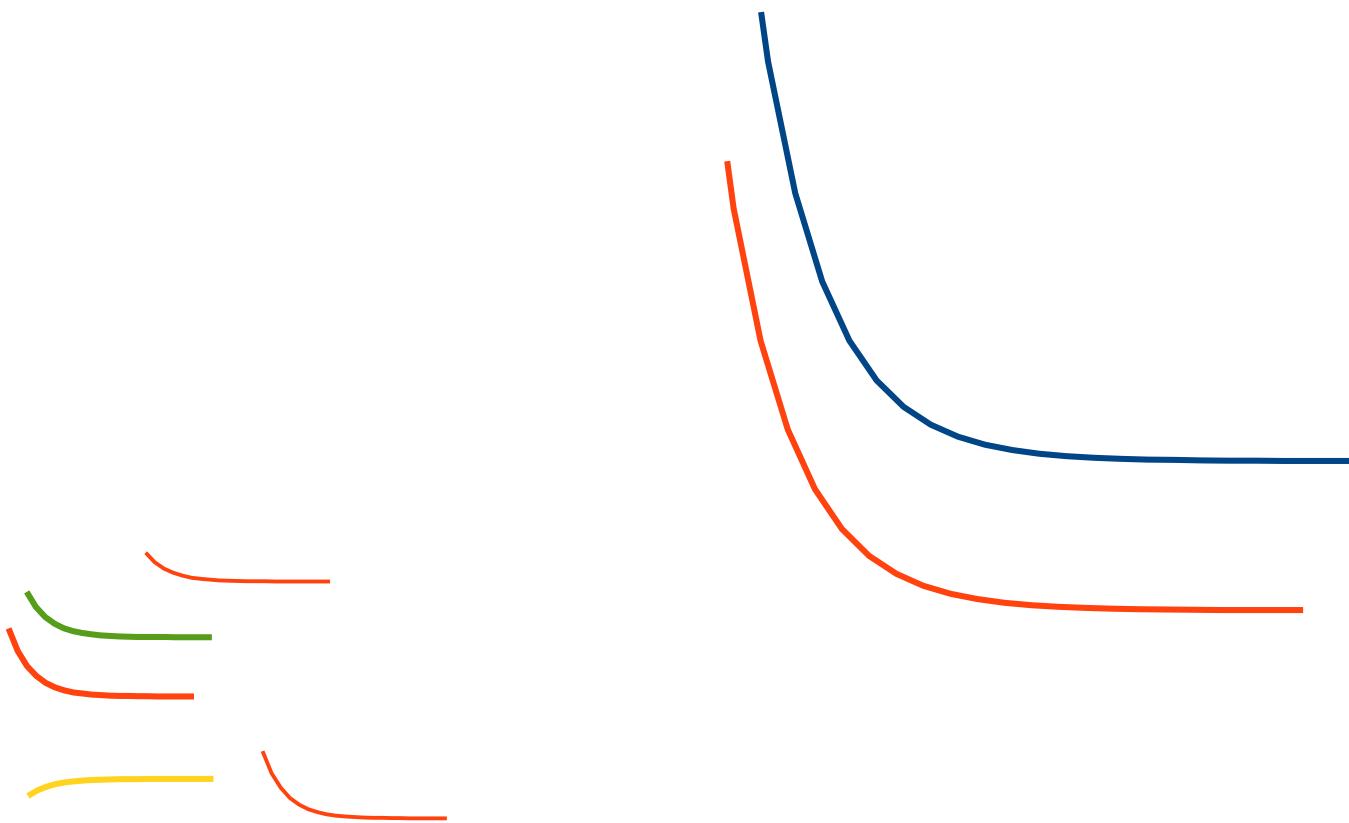
$$K_2 = \frac{1}{2} \sqrt{A^2+B^2} e^{-j\theta} G(-j\omega)$$

Frequency Response in Control Theory (4)



$$\begin{aligned}
 Y_{ss}(s) &= \frac{K_1}{s+j\omega} + \frac{K_2}{s-j\omega} \\
 &= \frac{1}{2}(A+jB)G(-j\omega)\frac{1}{s+j\omega} + \frac{1}{2}(A-jB)G(+j\omega)\frac{1}{s-j\omega} \\
 &= \frac{1}{2}\sqrt{A^2+B^2}e^{+j\theta}G(-j\omega)\frac{1}{s+j\omega} + \frac{1}{2}\sqrt{A^2+B^2}e^{-j\theta}G(+j\omega)\frac{1}{s-j\omega} \\
 \\[10pt]
 y_{ss}(t) &= \frac{\sqrt{A^2+B^2}}{2}[G(-j\omega)e^{-j\omega t}e^{+j\theta} + G(+j\omega)e^{+j\omega t}e^{-j\theta}] \\
 &= \frac{\sqrt{A^2+B^2}}{2}[G(+j\omega)e^{-j(\omega t-\theta)} + G(+j\omega)e^{+j(\omega t-\theta)}] \\
 &= \sqrt{A^2+B^2} \Re\{G(+j\omega)e^{+j(\omega t-\theta)}\} \\
 &= \sqrt{A^2+B^2}|G(+j\omega)| \cos(\omega t - \theta + \arg\{G(+j\omega)\})
 \end{aligned}$$

Impulse Response $h(t)$



References

- [1] <http://en.wikipedia.org/>
- [2] J.H. McClellan, et al., Signal Processing First, Pearson Prentice Hall, 2003
- [3] M. J. Roberts, Fundamentals of Signals and Systems
- [4] S. J. Orfanidis, Introduction to Signal Processing
- [5] B. P. Lathi, Signals and Systems